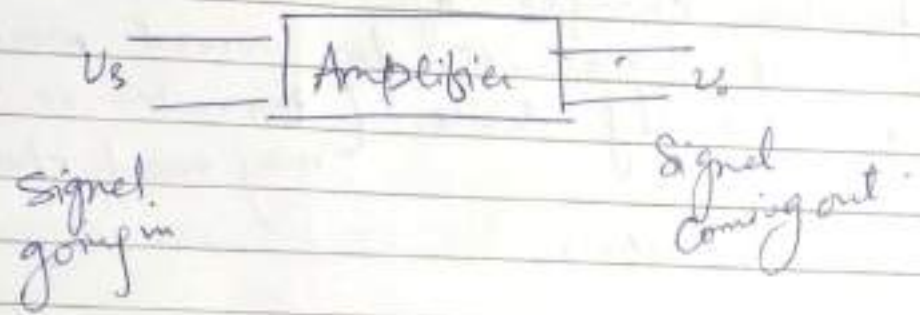


# Feedback in Amplifiers

Positive and Negative feedback, effect of negative feedback on I/P impedance, O/P impedance, Gain, stability, Distortion and noise.

What is feedback? What kind of feedbacks we can use?

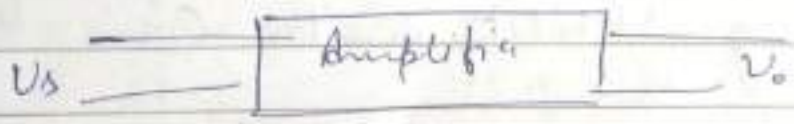
## FEEDBACK



Purpose of Amplifier is to amplify the signal without changing its characteristics except the magnitude.

∴ Output of Amplifier is a fun<sup>n</sup> of I/P.

If O/P is not influenced by I/P, The ckt is called open loop having no feedback i.e.



Here, no O/P is not fed back to I/P.

But this is not a good ckt ∴ performance of the amplifier will change if the situations such as temp<sup>n</sup> variation will change.

In an Open loop amplifier  
Voltage gain  $A = \frac{V_o}{V_s}$  or  $V_o = A V_s$

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If gain changes i.e. If  $A$  becomes  $A + \Delta A$   
then o/p will change i.e.  $V_o + \Delta V_o$

$\Delta V_o = \Delta A \cdot V_s$  any change in gain of amplifier will reflect change in o/p voltage which is not acceptable

Gain may change from

- ① Temperature changes of environment
- ② Aging (Use of a ckt for several months/years)
- ③ Replacement of device. (transistor or resistors may need change)
- ④ Voltage fluctuations

We are supposed to design a ckt which will not be affected by above mentioned changes in Gain. This is achieved by feedback.

feedback  $\rightarrow$  A fraction of o/p <sup>voltage</sup> mixed with I/P.

$\therefore$  Real I/P is Geometrical sum of applied signal and the fraction which we are feeding back. This will bring stability in the Gain.

The types of Feedbacks :-

- ① ~~Positive feedback~~ Input =  $V_s + V_f$  where  $V_f$  = Return signal or fraction of o/p being fed to input.  
Hence, o/p will also increase and these steps cumulate and ultimately the gain will become very high and lead to "Instability" to the ckt. This is known as positive feedback. Positive feedback is



avoided in Amplifiers, but used in oscillator circuits.

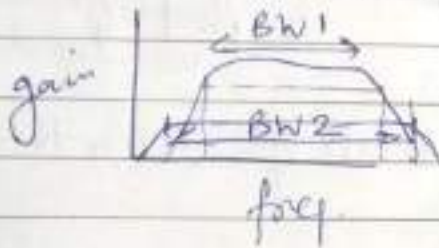
② Negative feedback

off phase is opposite to I/P phase in -ve feedback.

Net I/P to the amplifier is  $= V_s - V_f$

-ve feedback can modify ckt. performance such as

- ① Stability of gain against all source of variations
- ② Reduces non-linear distortions
- ③ Increases the Bandwidth (BW)



$BW1 < BW2$

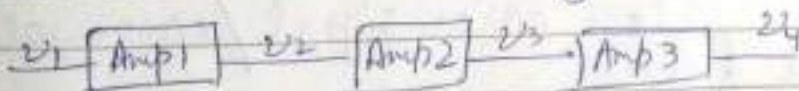
$BW1 \rightarrow$  without -ve feedback

$BW2 \rightarrow$  with -ve feedback

④ -ve feedback modifies (increase or decrease) the I/P and o/p impedances

One has to pay the price for all these advantages i.e. the disadvantage of negative feedback is that it reduces the gain of the amplifier.

Hence, multistage amplifier is used to increase the overall gain amplifier i.e.



$A_1 = \frac{v_2}{v_1}$        $A_2 = \frac{v_3}{v_2}$        $A_3 = \frac{v_4}{v_3}$

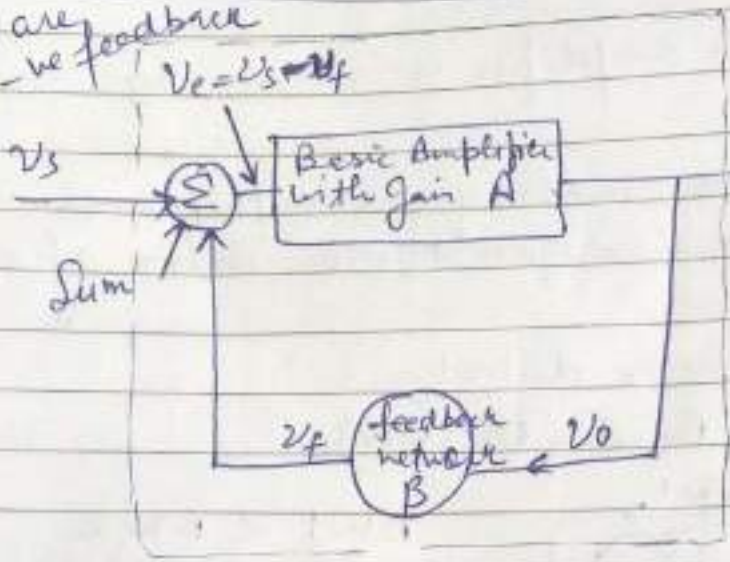
Overall gain of Amplifier  $A = \frac{v_4}{v_1} = A_1 \times A_2 \times A_3 = \frac{v_2}{v_1} \times \frac{v_3}{v_2} \times \frac{v_4}{v_3}$

i. Gain Reduction in Negative feedback is not a big problem. It can be boosted with by using Multistage Amplifiers.

General feedback Relation:

A = Gain of Basic Amplifier  
β = Gain of feedback network

Here we are considering -ve feedback



Amplifier with -ve feedback

Feedback network usually consists of one or two resistors

Various Quantities involved.

A = Open loop gain of amplifier  
ie. gain without feedback

β = feedback factor or Gain of feedback circuit  
ie.  $\beta = \frac{\text{Output voltage}}{\text{I/P voltage}}$

ie.  $\beta = \frac{V_f}{V_o}$

where  $V_f$  = feedback voltage  
 $V_o$  = o/p

$V_e$  = error signal / error voltage

From above figure

- $V_o = A V_e$  (1)
- $V_e = V_s - V_f$  (2)
- $V_f = \beta V_o$  (3)



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Substitute  $V_e$  from (3) into (2)

$$\text{i.e. } V_e = V_s - \beta V_o$$

Substitute this value of  $V_e$  into Eq<sup>n</sup> (1)

$$V_o = A(V_s - \beta V_o)$$

$$V_o(1 + \beta A) = A V_s$$

Gain of Amplifier with feedback  $A_{FB} = \frac{V_o}{V_s} = \frac{A}{1 + \beta A}$

$$A_{FB} = \frac{A}{1 + \beta A}$$

Gain with feedback  
for Negative feedback

$$A_{FB} = \frac{A}{1 - \beta A} \text{ for Positive feedback}$$

$\beta A \rightarrow$  called as "Loop Gain"

① GAIN STABILITY  
(OP voltage do not change due to any ckt modifications).

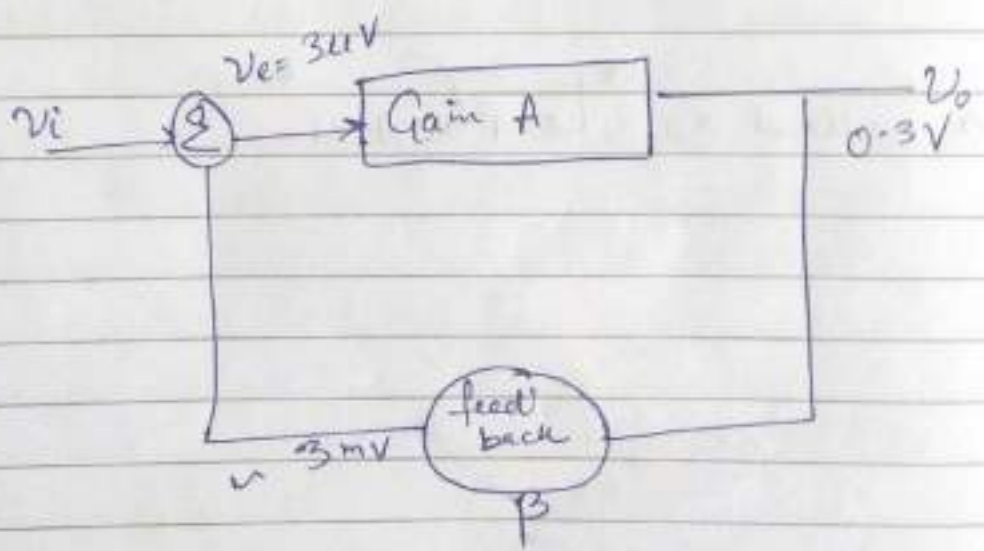
$$A_{fb} = \frac{A}{1 + \beta A}$$

If  $\beta A \gg 1$   
 $A_{fb} \approx \frac{A}{\beta A} = \frac{1}{\beta}$

i.e.  $A_{fb} = \frac{1}{\beta}$   $\therefore$  Gain of feedback amplifier is totally independent of A

$\therefore$  -ve feedback provides high stability to the feedback amplifier.

Example



- find
- Open loop Gain A
  - Closed loop Gain  $A_{fb}$
  - Gain of feedback circuit  $\beta$

$$\text{Sol}^n - A = \frac{0.3V}{3\mu V} = 10^5$$

$$\beta = \frac{V_f}{V_o} = \frac{3mV}{0.3V} = 10^{-2}$$

$$A_{fb} = \frac{A}{1 + \beta A} = \frac{10^5}{1 + (10^{-2})(10^5)} = \frac{10^5}{1 + 10^3} = 10^2 = 100$$

## ② Reduction in Non-linear distortion

Distortion finally comes in form of voltage fluctuation.

For Eg. - If a ckt incorporates distortion of the order of 10%, without feedback this means that if the o/p voltage is 10V, it will be 9V or 10V.

By incorporating -ve feedback it can be minimized to 1% or 2%.

Let's define Distortion as  $D = \frac{\text{change in gain}}{\text{gain}}$

Here D corresponds to open loop.

$$D = \frac{\text{change in gain}}{\text{gain}} = \frac{\Delta A}{A}$$

for small change  $\Delta A \rightarrow dA$

$$\therefore D = \frac{dA}{A} \quad (1)$$

In the same way  $D_{fb} = \frac{dA_{fb}}{A_{fb}} \quad (2)$



Rel<sup>n</sup> b/w D and Dfb

$$A_{fb} = \frac{A}{1+BA} \quad (1)$$

Differentiating wrt gain A, we get

$$dA_{fb} = \frac{(1+BA)dA - BA dA}{(1+BA)^2}$$

$$dA_{fb} = \frac{dA}{(1+BA)^2} = \frac{dA}{(1+BA)} \cdot \frac{1}{(1+BA)}$$

Since  $\frac{1}{1+BA} = \frac{A_{fb}}{A}$  from (1)

$$\therefore dA_{fb} = \frac{dA}{(1+BA)} \cdot \frac{A_{fb}}{A}$$

$$\frac{dA_{fb}}{A_{fb}} = \frac{dA}{A} \cdot \frac{1}{1+BA}$$

Using (1) & (2)

$$D_{fb} = \frac{D}{1+BA}$$

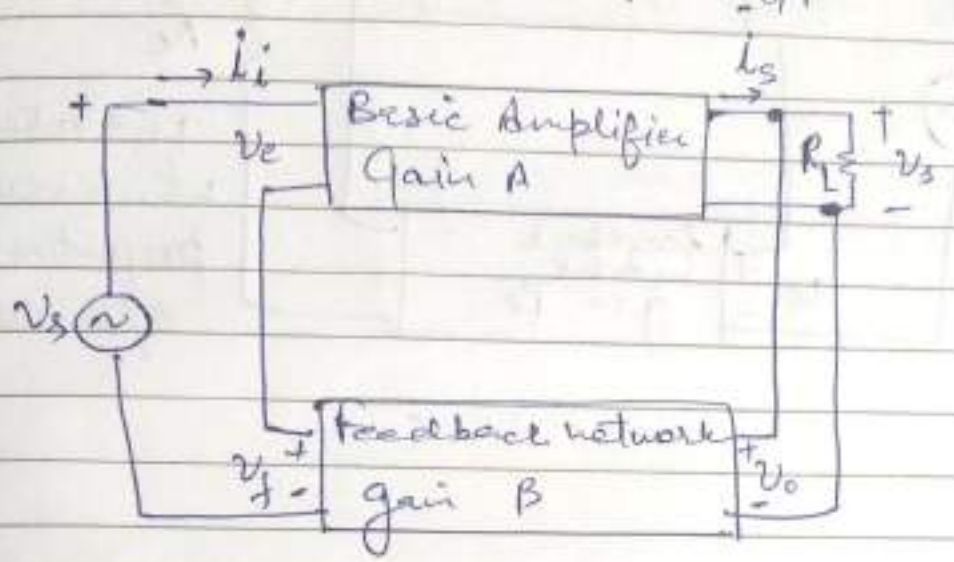
∴ Distortion reduces due to -ve feedback.



Different types of -ve feedback.

① <sup>o/p</sup> <sup>i/p</sup> Voltage Series feedback.

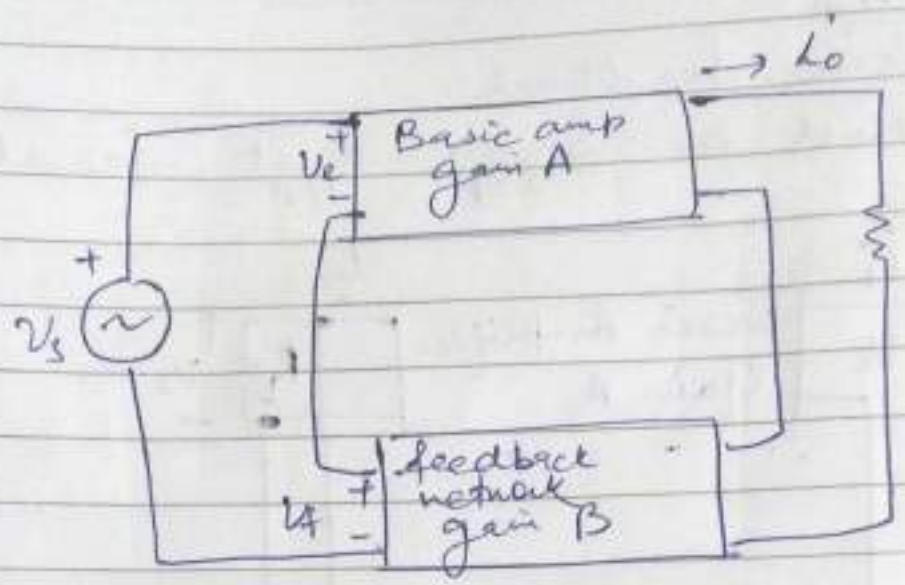
Also called as <sup>i/p</sup> Series - <sup>o/p</sup> Shunt feedback.



At the <sup>i/p</sup> the ckt has Series Connection  
At the <sup>o/p</sup> - - - - - Shunt (parallel) connection

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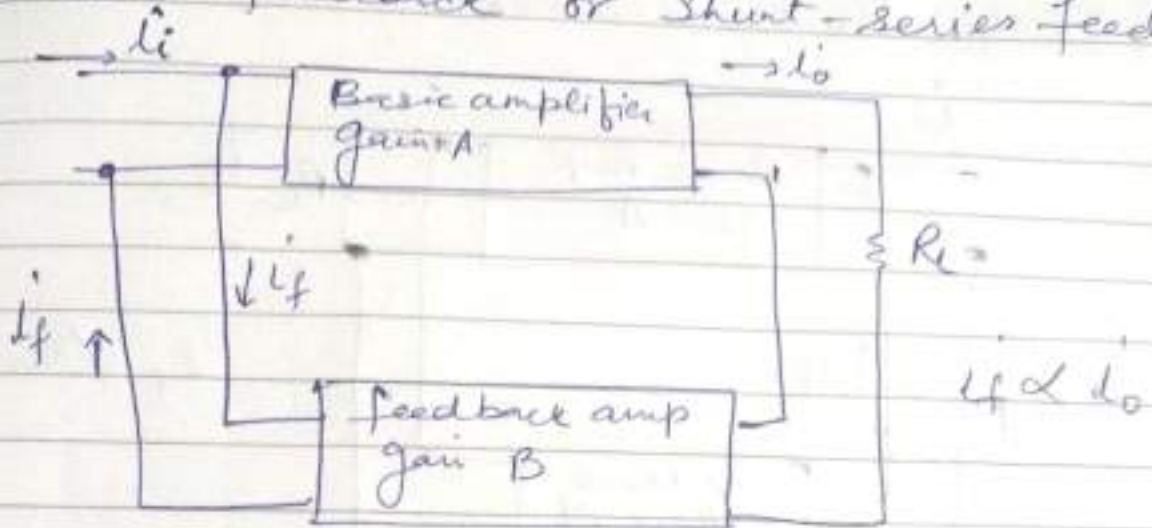
## ② Current Series feedback or Series-Series feedback



$v_o \rightarrow$  return signal which varies in proportion to the o/p current.

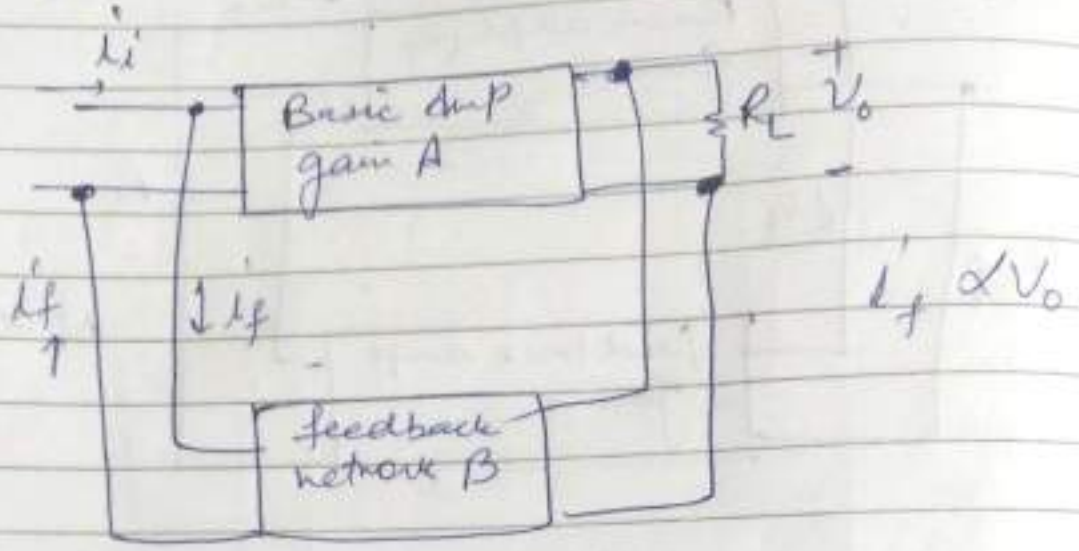


③ Current Shunt feedback or Shunt-series feedback.



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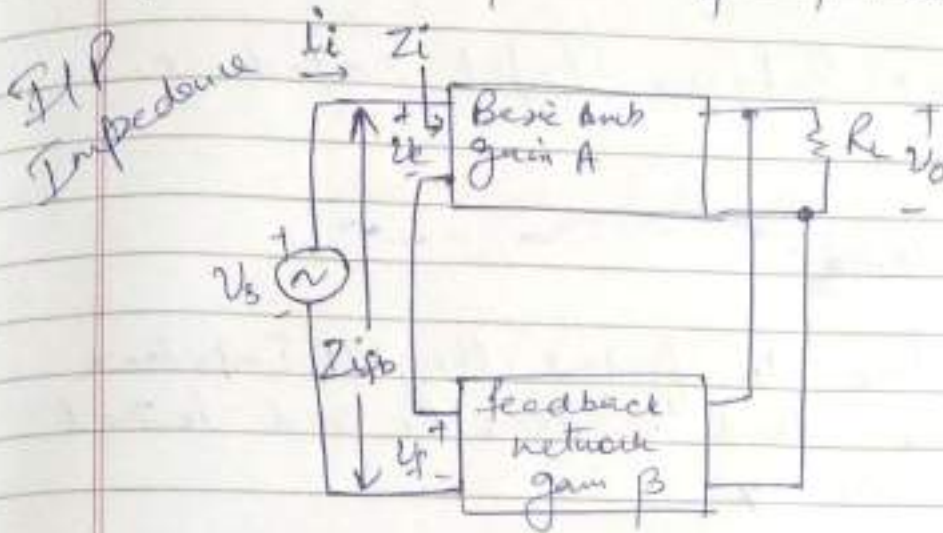
Voltage-Shunt feedback or Shunt-Shunt feedback





Effect of -ve feedback on Input and Output Impedances of an amplifier.

We are doing it for Voltage Series negative feedback Amplifier  
If the input impedance of open loop amplifier =  $Z_i$



$$\beta = \frac{V_f}{V_o}$$

Input Impedance for open loop amp  $Z_i = \frac{V_e}{I_i}$

For feedback ckt  $Z_{iFB} = \frac{V_s}{I_i}$

As  $V_e = V_s - V_f$  and  $\beta = \frac{V_f}{V_o}$

$\therefore V_f = \beta V_o$

$\therefore V_e = V_s - \beta V_o$

$V_s = V_e + \beta V_o$

Also  $A = \frac{V_o}{V_e}$

or  $V_o = A V_e$

$\therefore V_s = V_e + \beta A V_e$

$V_s = V_e (1 + \beta A)$

Dividing both sides by  $I_i$

$$\frac{V_s}{I_i} = \frac{V_e}{I_i} (1 + \beta A) \Rightarrow \boxed{Z_{iFB} = Z_i (1 + \beta A)}$$

$\therefore$  I/P impedance increases by a factor of  $(1 + \beta A)$  during -ve feedback.

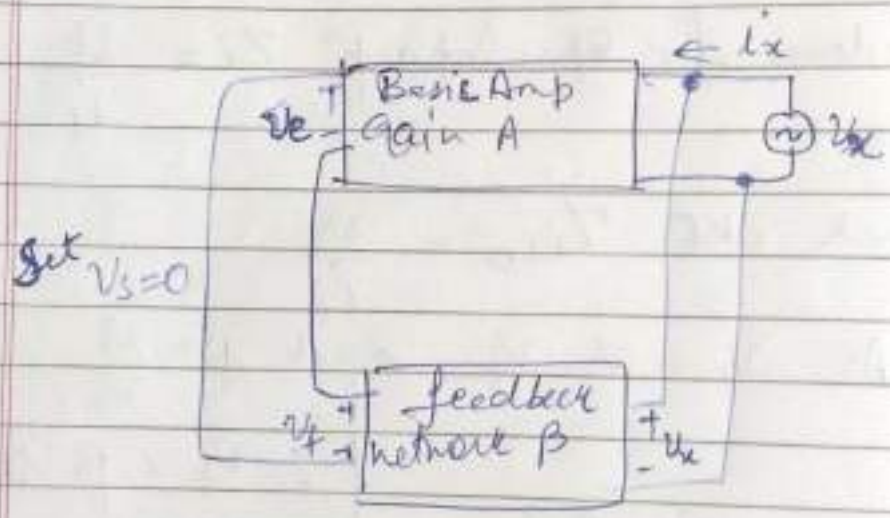
eg If  $Z_i = 1\text{ k}\Omega$   
Let  $\beta A = 100$

$$Z_{i_{fb}} \approx 100\text{ k}\Omega$$

Negative feedback Enhances Input Impedance.

### Output Impedance

Whenever we have to find out the O/P Impedance we have to remove load resistance  $R_L$  and instead add voltage source  $V_x$



$$Z_{o_{fb}} = \frac{V_x}{i_x}$$

$$V_e = V_s - V_f$$

Since  $V_s = 0$

$$V_e = -V_f$$

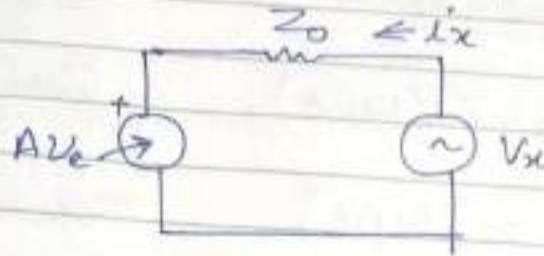
$$\therefore V_e = -\beta V_x$$

$$\beta = \frac{V_f}{V_x}$$

$$V_f = \beta V_e$$



from the equivalent circuit for the output port



$$i_x = \frac{V_x - A V_e}{Z_o}$$

$$i_x = \frac{V_x + \beta A V_x}{Z_o} \quad \text{Because } V_e = -\beta V_x$$

$$i_x = \frac{V_x (1 + \beta A)}{Z_o}$$

$$Z_{o\text{fb}} = \frac{V_x}{i_x} = \frac{Z_o}{1 + \beta A}$$

$$Z_{o\text{fb}} = \frac{Z_o}{1 + \beta A}$$

O/P impedance reduces by the factor of  $\frac{1}{1 + \beta A}$

for other 3 types of feedbacks  $Z_{i\text{fb}}$  and  $Z_{o\text{fb}}$  has to be found out by yourself:

$Z_i = I/P$  Impedance without feedback  
 $Z_o = O/P$

<u>Feedback Configuration</u>	<u>I/P Impedance</u>	<u>O/P Impedance</u>
① Voltage - Series FB	$Z_{ifs} = Z_i (1 + \beta A)$	$Z_{ofs} = \frac{Z_o}{1 + \beta A}$
② Current - Series FB	$Z_{ifs} = Z_i (1 + \beta A)$	$Z_{ofs} = \frac{Z_o (1 + \beta A)}{1 + \beta A}$
③ Voltage Shunt FB	$Z_{ifs} = \frac{Z_i}{1 + \beta A}$	$Z_{ofs} = \frac{Z_o}{1 + \beta A}$
④ Current - Shunt	$Z_{ifs} = \frac{Z_i}{1 + \beta A}$	$Z_{ofs} = Z_o (1 + \beta A)$

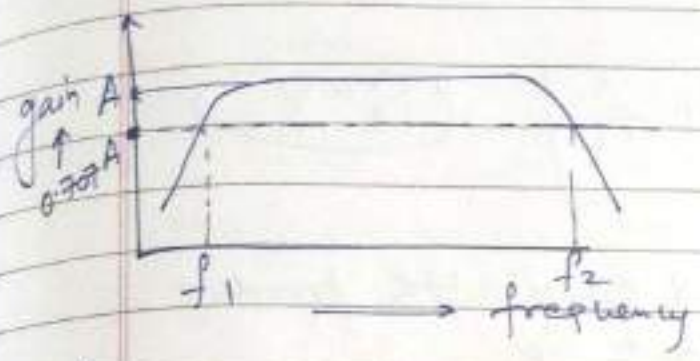
Depending upon the requirement apart from Stability, any one of the configurations with required impedances could be used.

When the connection of I/P and feedback network is in series then the resistance will increase.

When the connection feedback network is connected in parallel then the resistance will decrease.

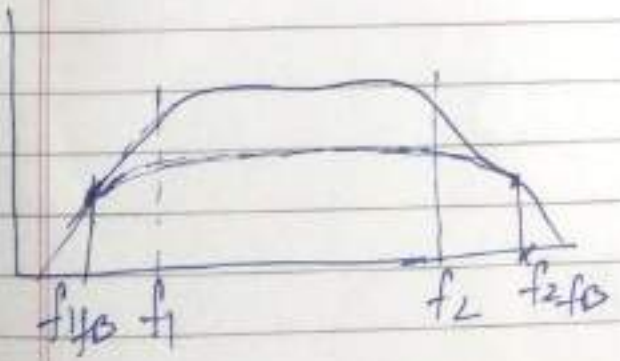


③ Effect of -ve feedback on Bandwidth of the amplifier.



$f_1$  → lower Cut off frequency without feedback  
 $f_2$  → Upper Cut off frequency " "  
 $f_2 - f_1 = \text{Bandwidth}$

Effect of -ve feedback is to resist any change in circuit performance.  
 $\therefore f_{1fb}$  will lie to the left of  $f_1$  (without feedback)  
 and  $f_{2fb}$  " " " " right of  $f_2$  ( " )



$f_{1fb} = \frac{f_1}{1+BA}$   
 i.e.  $f_{1fb}$  reduces by a factor of  $\frac{1}{1+BA}$ .

and  $f_{2fb} = (1+BA)f_2$   
 i.e.  $f_{2fb}$  enhances by a factor of  $(1+BA)$

Often  $f_2 \gg f_1$



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Bandwidth  $BW = f_2 - f_1$   
 Since  $f_2 \gg f_1$   
 $BW = f_2$

or  $f_2 f_0 = (1 + \beta A) f_2$   $E_4 \times$

Proof:-

for all amplifiers Gain  $\times$  Bandwidth product is constant i.e.

Gain  $\times$  Bandwidth = Constant =  $f_T$   $E_4 \times$

$f_T$  = upper cutoff frequency at which gain falls to unity

We can write  $E_4 \times$  for any frequency

If  $A \rightarrow$  Gain of Internal/basic amplifier

$A \cdot f_2 = f_T$

for feedback

$A_{f_0} f_2 f_0 = f_T$

Since  $A_{f_0} = \frac{A}{1 + \beta A}$

$\left( \frac{A}{1 + \beta A} \right) f_2 f_0 = f_T$

or  $\left( \frac{A}{1 + \beta A} \right) f_2 f_0 = A f_2$

Hence  $f_2 f_0 = (1 + \beta A) f_2$

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Eg If  $H\beta A = 50$

$$A = 10000$$

$$A_{\beta B} = \frac{10000}{50} = 200$$

But Bandwidth will increase due to negative feedback.

If  $f_2 = 10 \text{ kHz}$

$$f_{2\beta B} = (1 + \beta A) f_2$$

$$= 50 \times 10 \text{ kHz}$$

$$= 500 \text{ kHz}$$

Only thing we sacrifice using negative feedback is gain.  $(1 + \beta A) \rightarrow$  Sacrifice factor.